

① LINEAR SYSTEMS THEORY

GOAL: provide a concise way of characterising

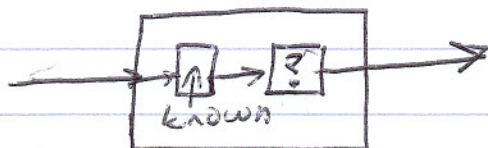


so that one can (a) describe it

(b) test models for it

(c) make guesses as to what's inside

(d) make use of partial knowledge



Examples electric circuit

physiology

VOR - rotate head, observe eye movement

volume homeostasis - apply volume load

membrane . urine output adjust

- apply current, measure voltage

light to voltage

neural activity to blood flow

For general F , this is intractable.

So we make 2 assumptions:

translation-invariance

linearity

(+ causality)

(Post-processing filters need not be causal.)

②

Translation-invariance: if $[F(s)](t) = r(t)$

and

$$s' = D_\tau(s) \quad (\text{i.e., } s'(t) = s(t+\tau))$$

then

$$[F(s')](t) = r(t+\tau)$$

$$\text{i.e., } F D_\tau = D_\tau F$$

Linearity: $[F(s_1 + s_2)](t) = [F(s_1)](t) + [F(s_2)](t)$

and

$$[F(\alpha s)](t) = \alpha [F(s)](t)$$

Linearity is always an approximation.

The theory does extend to non-linear transformations
in part.

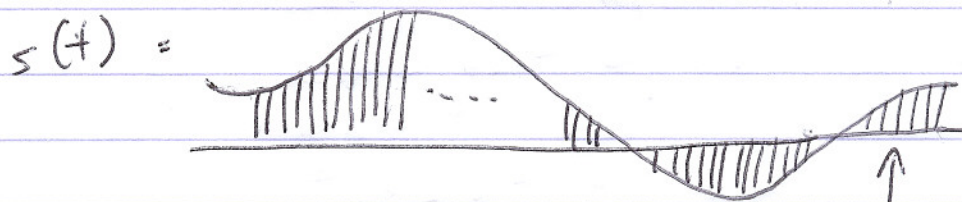
Linearity allows us to consider F as an element of $\text{Hom}(V, V)$
where V is the V.S. of functions of time.

But if F is a physical system, then how to interpret
 $[F(s)](t)$ for s complex-valued?

Ans: linearity $s(t) = a(t) + i b(t)$
so $F(s)(t) = [F(a)](t) + i [F(b)](t)$

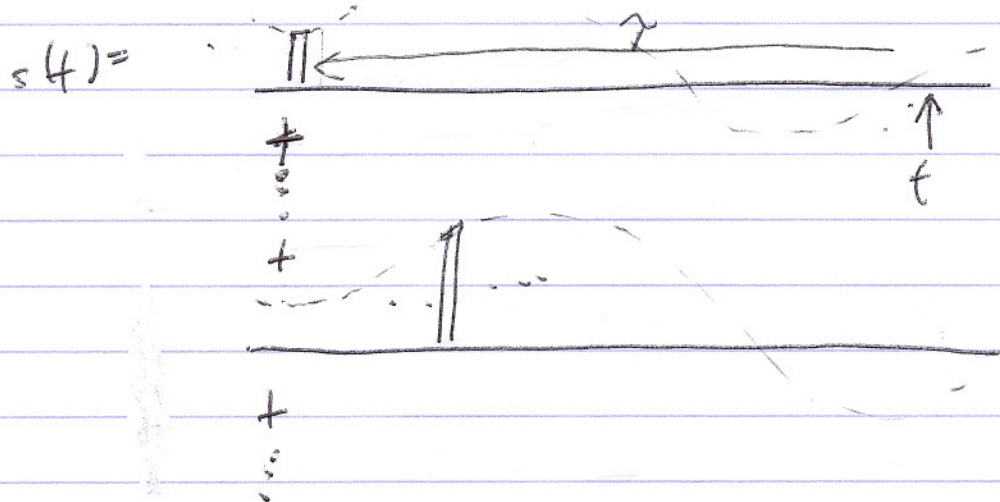
③

Linearity gives us an intuitive way of characterizing F .



Parcel it into small intervals

Linearity means that we can consider each interval independently.



Response at time $t =$ sum of contributions of s at all times in past, $t - \tau$.

Say a unit pulse at time 0 leads to a response $f(\tau)$.
Equivalently, a unit pulse at time $t - \tau$ leads to a response $f(\tau)$ at time t .

$$r(t) = \int_{\tau=0}^{\infty} f(\tau) s(t - \tau) d\tau.$$

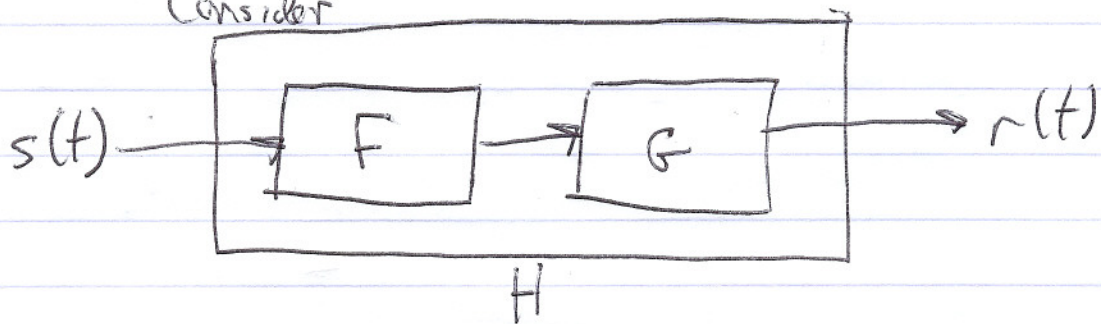
f is the "impulse response"

④

What's the problem with this description?

Ⓐ Estimating $f(\tau)$. You need to work where linearity is most likely to break down.

Ⓑ Combining systems. Parallel is easy.
Consider



If you know $f(\tau)$ & $g(\tau)$, what's $h(\tau)$?

say $F(s) = q$, $G(q) = r$.

$$r(t) = \int g(\tau) q(t - \tau) d\tau$$

$$q(t) = \int f(\tau') s(t - \tau') d\tau'$$

$$\text{so } r(t) = \iint g(\tau) f(\tau') s(t - \tau - \tau') d\tau' d\tau$$

$$u = \tau + \tau'$$

$$= \iint g(\tau) f(u - \tau) s(t - u) d\tau du$$

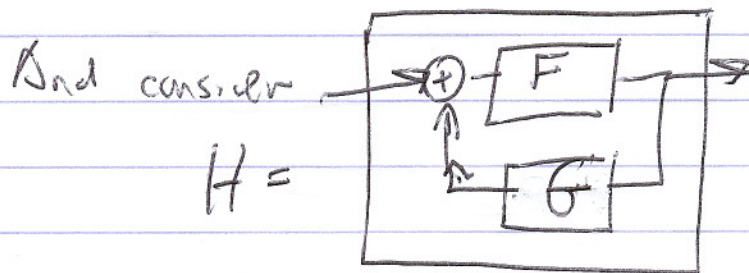
$$= \int h(u) s(t - u) du \quad \text{for } h(u) = \int g(\tau) f(u - \tau) d\tau$$

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So, for systems in series,

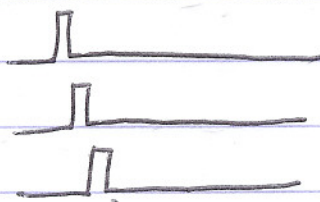
$$h(u) = \int g(\tau) f(u-\tau) d\tau, \text{ or } h = f * g$$

It's not so easy to intuit what this means, or, say, given h and f , to solve for g .



or more complex networks

What we've done above is to describe F in terms of the basis set of time functions



These are the delta-functions $\delta(t-\tau)$

D_T permutes them. But we know that there is another basis set for which D_T multiplies them, by $e^{i\omega t}$.

So now let's change bases. In the new basis set, $e^{i\omega t}$, F must act as multiplication. i.e., for $s(t) = e^{i\omega t}$, then $[F(s)](t) =$ a multiple of $e^{i\omega t}$.

⑥

Let's find two multiple, call it $\hat{f}(\omega)$, the "transfer function" of f .
 If $s(t) = e^{i\omega t}$ then

$$\begin{aligned} [F(s)](t) &= \int f(\tau) s(t-\tau) d\tau \\ &= \int f(\tau) e^{i\omega(t-\tau)} d\tau \\ &= e^{i\omega t} \int f(\tau) e^{-i\omega\tau} d\tau \end{aligned}$$

$$\text{So, } \boxed{\hat{f}(\omega) = \int e^{-i\omega\tau} f(\tau) d\tau}$$

Interpret $\hat{f}(\omega)$ in terms of real signals, say $\hat{f}(\omega) = |f(\omega)| e^{i\phi(\omega)}$

$$s(t) = e^{i\omega t} = \cos(\omega t) + i \sin(\omega t)$$

$$r(t) = |f(\omega)| e^{i\phi(\omega)} s(t)$$

$$= |f(\omega)| (\cos \phi(\omega) + i \sin \phi(\omega)) (\cos \omega t + i \sin \omega t)$$

$$= |f(\omega)| \left[\begin{aligned} &[\cos \phi(\omega) \cos \omega t - \sin \phi(\omega) \sin \omega t] \\ &+ i [\sin \phi(\omega) \cos \omega t + \cos \phi(\omega) \sin \omega t] \end{aligned} \right]$$

$$= |f(\omega)| \left[\cos(\omega t + \phi(\omega)) + i \sin(\omega t + \phi(\omega)) \right]$$

\uparrow
Amplitude

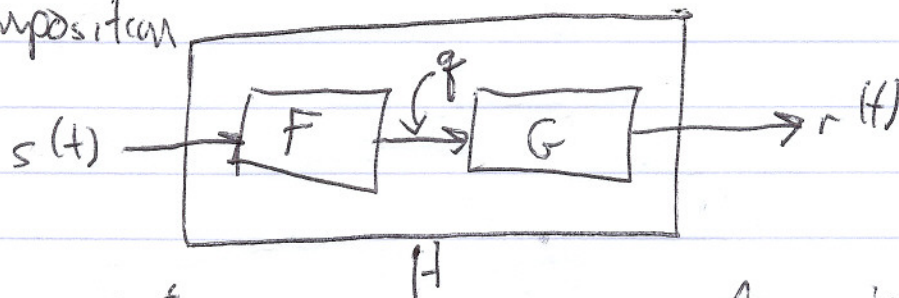
\uparrow
phase shift

⑦

Does the transfer function help with the "problems" on p. 4?

Ⓐ measurement. Use $\cos(\omega t)$.

Ⓑ composition



Put in $e^{i\omega t} = s(t)$. Then $q(t) = \hat{f}(\omega) e^{i\omega t}$.

$$\begin{aligned} \text{Next, since } G \text{ is linear, } r(t) &= \hat{g}(\omega) [\hat{f}(\omega) e^{i\omega t}] \\ &= \hat{g}(\omega) \hat{f}(\omega) e^{i\omega t} \end{aligned}$$

$$\text{So } \hat{h}(\omega) = \hat{f}(\omega) \hat{g}(\omega). \quad \text{"No Convolution Thm!"}$$

What if we want to calculate how F responds to arbitrary s , but only have $\hat{f}(\omega)$? i.e., can we find $f(t)$ from $\hat{f}(\omega)$?

$$\text{We need to write } f(t) = \int e^{i\omega t} \hat{f}(\omega) d\omega$$

Since, then we could calculate $f(t) = [F(f)](t)$ from its responses to $e^{i\omega t}$.

⑧

"It turns out" that

$$f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{i\omega t} d\omega$$

so the response to a δ -function is

$$f(t) = [F(\delta)](t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \hat{f}(\omega) e^{i\omega t} d\omega.$$

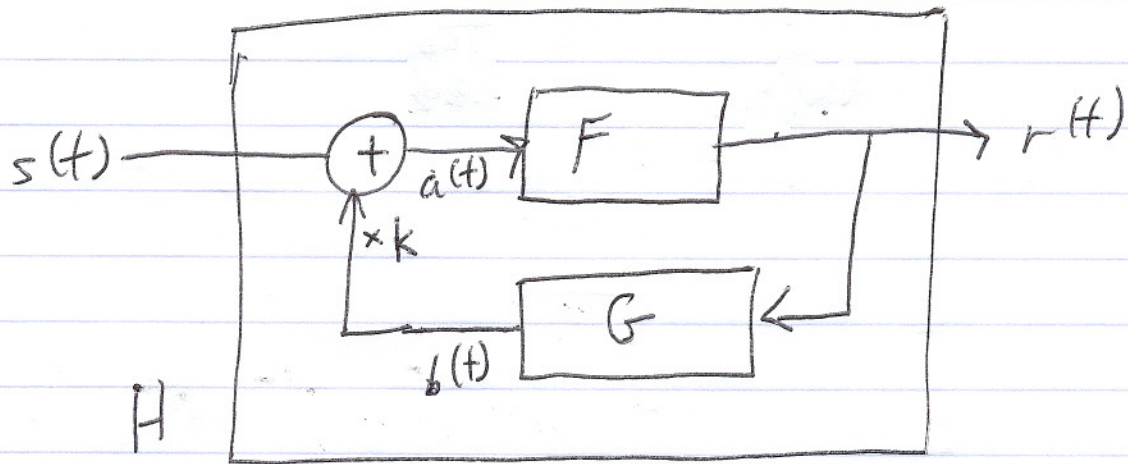
$$\begin{aligned} \hat{f}(\omega) &= \int_{-\infty}^{\infty} e^{-i\omega t} f(t) dt \\ f(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} e^{i\omega t} \hat{f}(\omega) d\omega \end{aligned}$$

Fourier Transform PAIRS

A way to see this - pp 4-5 of 2004 notes FAAP.

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An example of combining elements in a more complex way: feedback



(calculate $\hat{h}(\omega)$) say $s(t) = e^{i\omega t}$
 $r(t) = \hat{h}(\omega) e^{i\omega t}$

$$b(t) = \hat{g}(\omega) \hat{h}(\omega) e^{i\omega t}$$

$$a(t) = s(t) + kb(t) = e^{i\omega t} + k\hat{g}(\omega)\hat{h}(\omega)e^{i\omega t}$$

$$= \underbrace{e^{i\omega t} (1 + k\hat{g}(\omega)\hat{h}(\omega))}_{a(t)}$$

$$r(t) = [F(a)](t), \text{ so}$$

$$\hat{h}(\omega) e^{i\omega t} = \hat{f}(\omega) e^{i\omega t} (1 + k\hat{g}(\omega)\hat{h}(\omega))$$

$$\hat{h}(\omega) = \hat{f}(\omega) (1 + k\hat{g}(\omega)\hat{h}(\omega))$$

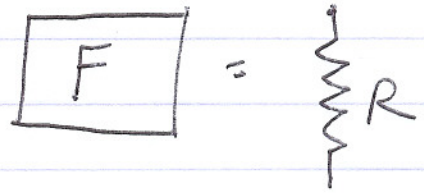
$$\hat{h}(\omega) = \frac{\hat{f}(\omega)}{1 - k\hat{f}(\omega)\hat{g}(\omega)}$$

(10)

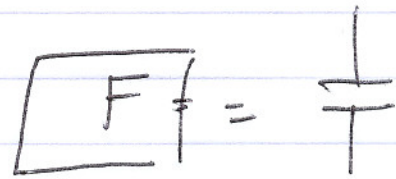
Apply current, measure voltage - in a network of R's - C's.

$$V = IR$$

$$\hat{f}(\omega) = R$$



$$Q = CV$$



$$I = \frac{dQ}{dt} = C \frac{dV}{dt}$$

$$\frac{dV}{dt} = \frac{1}{C} I$$

$$I = e^{i\omega t}, \quad V = \hat{f}(\omega) e^{i\omega t}$$

$$\Rightarrow \frac{dV}{dt} = i\omega e^{i\omega t} \hat{f}(\omega)$$

$$\hat{f}(\omega) i\omega e^{i\omega t} = \frac{1}{C} e^{i\omega t}$$

$$\hat{f}(\omega) = \frac{1}{i\omega C}$$

R's + C's will always lead to algebraic combination of $R = \frac{1}{i\omega C}$,
i.e., rational expressions in ω .