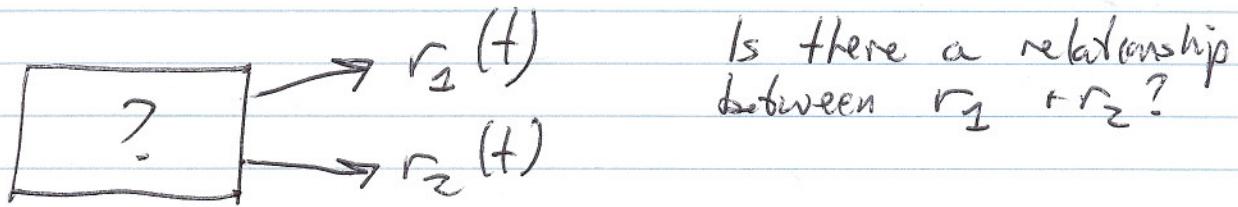
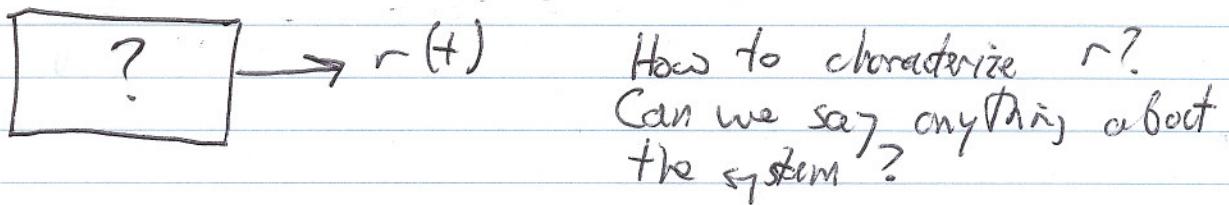
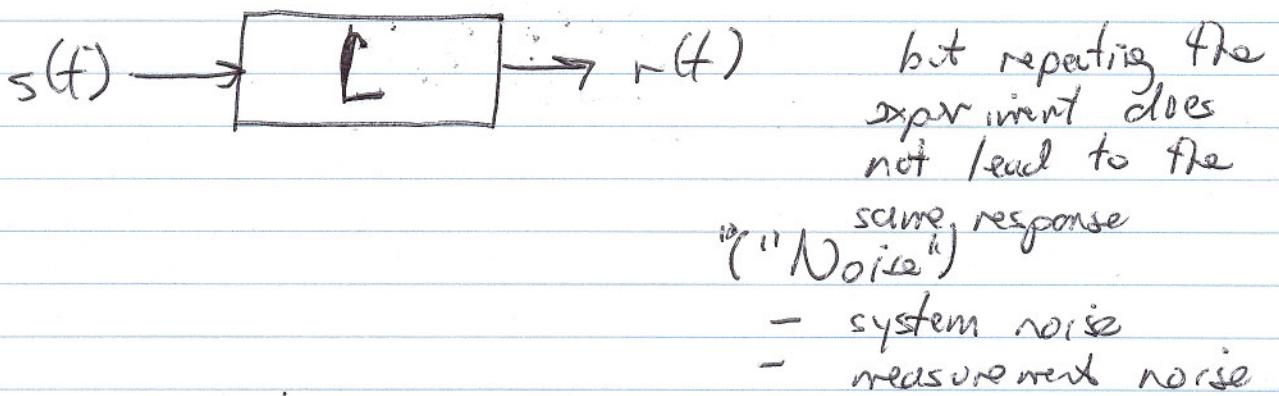


⑥ Fourier Analysis - Applications: Noise + Variability

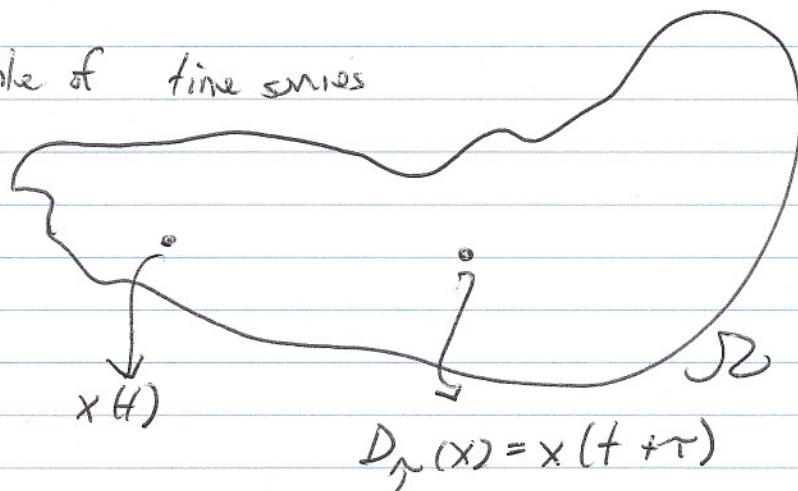
A cluster of settings + problems:



In each case, we need to make use of observations made at different times. But we're thinking of this as repeated observations of samples drawn from an ensemble.

(2)

An ensemble of time series

Characteristics $p(x)$

$$D_\tau(x) = x(t+\tau)$$

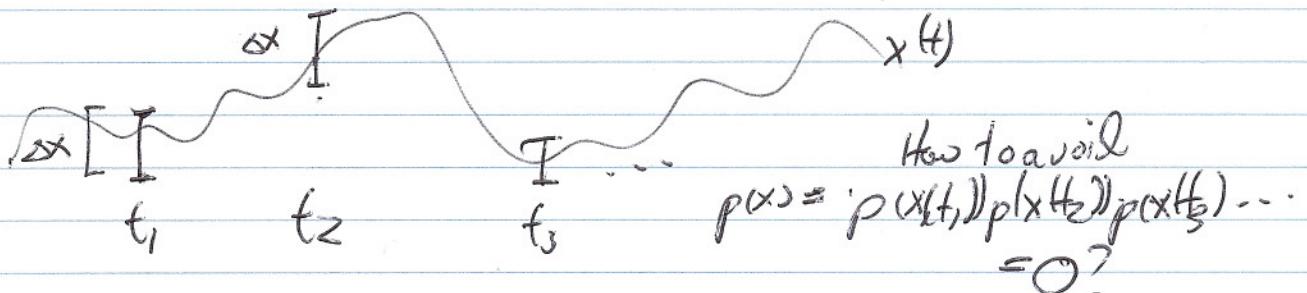
Typically, we "sample the ensemble" by waiting, resampling again.

So we need:

(a) Stationarity: $p(x(t)) = p(D_\tau(x)(t))$ (b) Ergodicity $\langle \dots \rangle_L = \langle \dots \rangle_T$

ensemble-average = time-average

(c) Limited correlations in time:

for sufficiently long T , $x(t)$ and $x(t+T)$
are independent.Note: How to define $p(x)$, when x is a time series?Easy to define $p(x(t_i))$, for t_i : a single time point

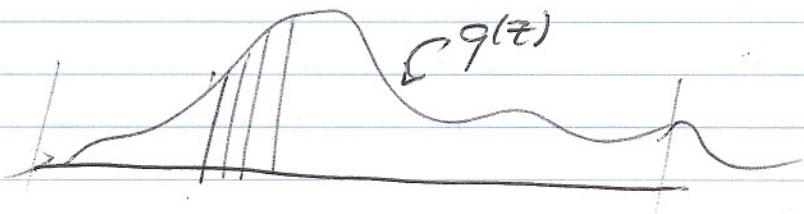
(2)

Probability $f(x) \propto f(t)$ jumps through hoops of width Δx , at N times,
is (for small Δx) proportional to $(\Delta x)^N$.

As $N \rightarrow \infty$, this goes to 0, but, this is not a problem...

it is analogous to what happens when you discretize a probability distribution, but (discrete bins of size Δx), but here, the distribution has N dimensions, & $N \rightarrow \infty$.

E.g.: say z is an ordinary (scalar) random variable:



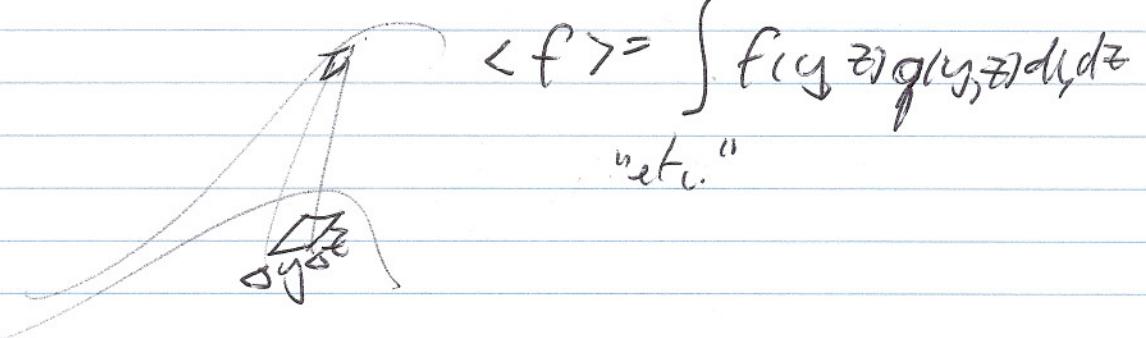
$g(z)$ is the probability dist. associated w/ z .

$g(z) \Delta z$ is prob that z is between z_0 & $z_0 + \Delta z$.

$$\text{Average } \langle z \rangle = \lim_{\Delta z \rightarrow 0} \sum z_i g(z_i) \Delta z = \int_{-\infty}^{\infty} z g(z) dz$$

$$\text{In general } \langle f(z) \rangle = \int_{-\infty}^{\infty} f(z) g(z) dz.$$

In 2-d,



$$\langle f \rangle = \int f(y, z) g(y, z) dy dz$$

"etc."

⑥

One way to understand $p(x)$ is from its moments
(mean, variance, covariance, etc.)

$\langle x(\tau) \rangle$ must be independent of τ (translation-invariance)

so it is not very useful. In fact, we always assume $\langle x(\tau) \rangle = 0$

So let's look at second moments, e.g.

$$\langle x(\tau_1) x(\tau_2) \rangle \text{ and } \langle x(\tau_1)^2 \rangle.$$

$$\text{Write } c_x(\tau_1, \tau_2) = \langle x(\tau_1) x(\tau_2) \rangle$$

$$(\langle x(\tau) \rangle \neq 0: c_x(\tau_1, \tau_2) = \langle (x(\tau_1) - \langle x(\tau) \rangle)(x(\tau_2) - \langle x(\tau) \rangle) \rangle)$$

$$\langle x(\tau_1 + t) x(\tau_2 + t) \rangle = \langle x(\tau_1) x(\tau_2) \rangle \quad (\text{Translation-invariance})$$

$$\therefore c_x(\tau_1 + t, \tau_2 + t) = c_x(\tau_1, \tau_2) \equiv c_x(\tau_1 - \tau_2)$$

This is the "auto-covariance". Note $c_x(\tau) = c_x(-\tau)$

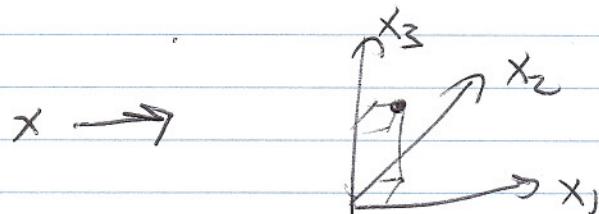
and

$$c_x(\tau) \leq c_x(0)$$

$$\text{Auto-correlation} = c_x(\tau) / c_x(0)$$

Geometrically: Sample $x(t)$ at τ_1, \dots, τ_n .

So $x(t)$ represented by an n -vector, $x(t, \cdot)$ ($i=1, \dots, n$)



distribution of x is a cloud.
What is its shape?

(5)

With n sample, there are $\frac{n(n-1)}{2}$ covariances,

some of them must be cpl ($\langle x_n x_{n+k} \rangle$)

and they are not independent:

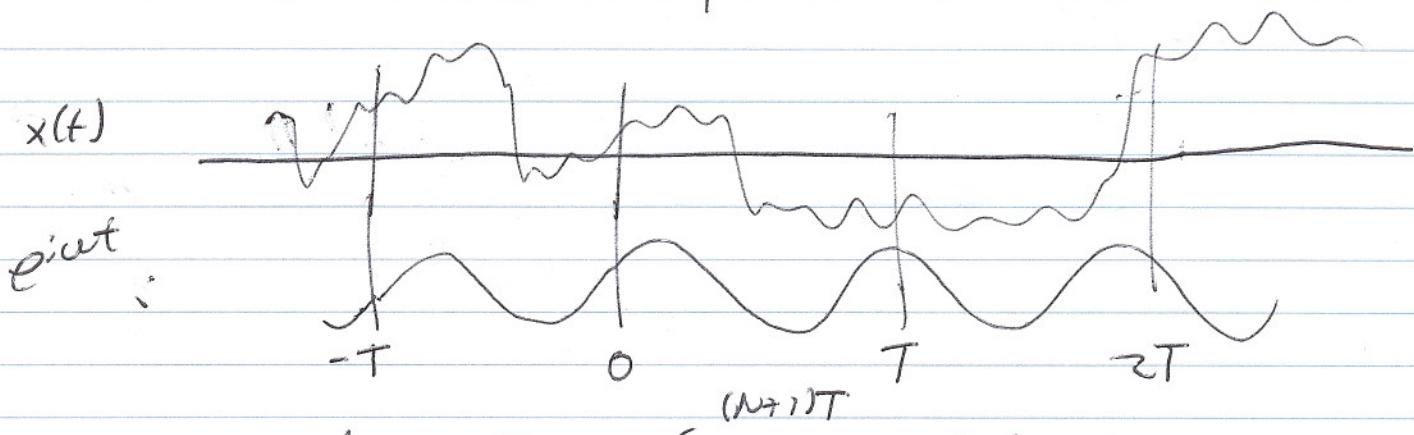
large $\langle x_2 x_3 \rangle$ and large $\langle x_3 x_5 \rangle$

\Rightarrow probably large $\langle x_2 x_5 \rangle$.

So how do we efficiently estimate $\hat{x}(t)$, &
how do we put error bars on it?

Change basis! Characterize $\tilde{x}(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$

Bst $\tilde{x}(\omega)$ is ill-defined



$$\int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \sum_{n=-\infty}^{\infty} x(nT) e^{-j\omega nT} = \sum_{n=-\infty}^{\infty} F_N$$

where F_N = a "Fourier estimate" at freq. ω , length T ,
start time $\frac{n}{NT}$

⑥

$$F_N = F(x; \omega, T, NT) \text{ where}$$

↑ ↑ ↑ ↑
signal freq. dur. start

$$F(x; \omega, T, T_0) = \int_{T_0}^{T_0+T} x(t) e^{-j\omega t} dt.$$

How is F distributed?

$$\begin{aligned} \text{Put } u &= t + T_0. \quad F(x; \omega, T, T_0) = \int_0^{T_0} x(u - T_0) e^{-j\omega(u-T_0)} du \\ &= e^{j\omega T_0} \int_0^T x(u - T_0) e^{-j\omega u} du \\ &= e^{j\omega T_0} \int_0^T [D_{-T_0}(x)](u) e^{-j\omega u} du \\ &= e^{j\omega T_0} F(D_{-T_0}(x); \omega, T, 0) \end{aligned}$$

B.t $F(D_{-T_0}(x); \omega, T, 0)$ must be distib like $F(x; \omega, T, 0)$
(translation-invariant),

Ans

$F(x; \omega, T, T_0)$ m.s.t have admstib indep of T_0 .

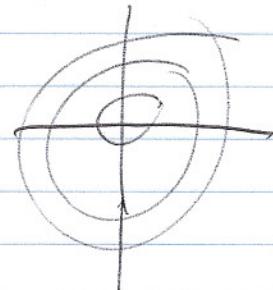
so $F(x; \omega, T, T_0)$ is circularly symmetric in the complex plane
Ans indep of T_0 .

⑦

$$I_{KT} = \int_0^{KT} x(t) e^{-j\omega t} dt = \sum_{N=0}^{K-1} F_N,$$

each F_N is a Fourier estimate, beginning at time

NT of duration T . H has a distribution
that is circularly symmetric
in the complex plane.

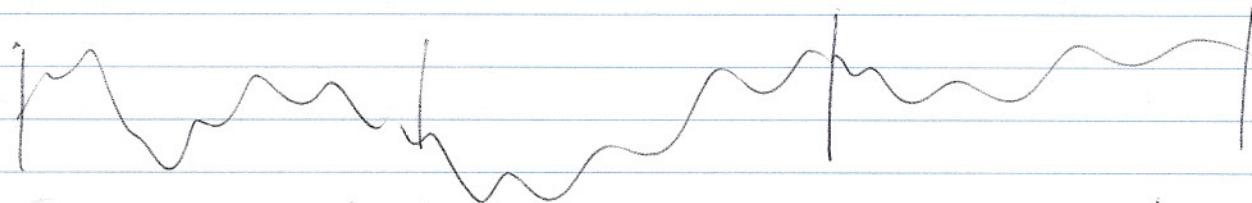


$$\text{so } I_{KT} = 0.$$

What about $\langle |I_{KT}|^2 \rangle$?

$$\begin{aligned} \langle |I_{KT}|^2 \rangle &= \left\langle \left| \sum_{N=0}^{K-1} F_N \right|^2 \right\rangle = \left\langle \sum_{N=0}^{K-1} F_N \sum_{M=0}^{K-1} \bar{F}_M \right\rangle \\ &= \sum_{N,M=0}^{K-1} \langle F_N \bar{F}_M \rangle. \end{aligned}$$

Two kinds of terms: $N=M$, $N \neq M$.



For long T , F_N and F_M are uncorrelated.

So, for long T , cross-terms $\rightarrow 0$ and

$$\langle |I_{KT}|^2 \rangle = \sum_{N=0}^{K-1} |F_N|^2 = K |\mathcal{F}(x, \omega, T, 0)|^2$$

(2)

$$I_{KT} = \int_0^{KT} x(t) e^{-i\omega t} dt$$

has a variance proportional to K , provided T is large enough. i.e.,

$$I_L = \int_0^L x(t) e^{-i\omega t} dt$$

a variance proportional to L (or a mean of 0)
for L sufficiently large.

$$P_x(\omega) = \lim_{L \rightarrow \infty} \frac{1}{L} \langle |I_L(\omega)|^2 \rangle$$

This is the "Power spectrum"

$$P_x(\omega) = \lim_{T \rightarrow \infty} \frac{1}{T} \langle \left| \int_0^T x(t) e^{-i\omega t} dt \right|^2 \rangle_T$$

Units of $P_x = X^2 \cdot \text{Time}$

Replacing $\langle \rangle_T$ by $\langle \rangle_{T_0}$,

$$P_x(\omega) = \lim_{T \rightarrow \infty} \frac{1}{N} \sum_{n=0}^{N-1} \frac{1}{T} |F(x_{\omega}, T, nT)|^2$$

Practical difficulties: Can't make $T \rightarrow \infty$.

With a finite amount of data, length L , must have $NT \leq L$.
So you can't have

$T \rightarrow \infty$

$N \rightarrow \infty$.

Large T : better est. of $F(x_{\omega}, T, nT)$ but fewer samples

⑦

What about the "off-diagonal" terms, like

$$\star \quad \langle F(x; \omega_1, L, T_1) F(x; \omega_2, L, T_2) \rangle ?$$

if start times & intervals $[T_1, T_1+L]$, $[T_2, T_2+L]$

are well-separated, then (c) \Rightarrow independence.

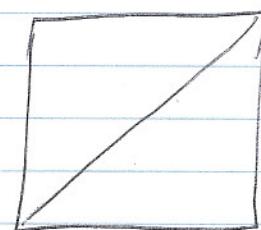
If not well-separated: say $T_1 = T_2 = T$

$$\langle \int_T^{L+T} x(t) e^{-i\omega_1 t} dt \int_T^{L+T} x(t) e^{-i\omega_2 t} dt \rangle$$

$$= \langle \int_T^{L+T} \int_T^{L+T} x(t_1) x(t_2) e^{-i\omega_1 t_1 - i\omega_2 t_2} dt_1 dt_2 \rangle$$

$$= \int_T^{L+T} \int_T^{L+T} \langle x(t_1) x(t_2) \rangle e^{-i\omega_1(t_1-t_2) - i\omega_2(t_1-t_2)} dt_1 dt_2$$

$$= \int_T^{L+T} \int_T^{L+T} c_{x(t_1-t_2)} e^{-i\omega_1(t_1-t_2) - i(\omega_1 + \omega_2)t_2} dt_1 dt_2$$



$$\gamma = t_1 - t_2$$

$$= \int_T^{L+T} \int \gamma c_x(\gamma) e^{-i\omega_1 \gamma} e^{-i(\omega_1 + \omega_2) t_2} d\gamma dt_2$$

$$t_2 = T - \gamma$$

$$\text{Put } u = t_2 - T.$$

$$= e^{-i(\omega_1 + \omega_2)T} \int_{u=0}^L \int \gamma c_x(\gamma) e^{-i\omega_1 \gamma} e^{-i(\omega_1 + \omega_2)u} d\gamma du.$$

⑥

This must (a) be indepd T + (b) be multiplied by
 $e^{-i(\omega_1 + \omega_2)T}$ as T advances. So, = 0 unless $\omega_1 + \omega_2 = 0$

$$\therefore \langle F(x; \omega_1, L, T) F(x; \omega_2, L, T) \rangle = 0$$

unless $\omega_1 + \omega_2 = 0$. if $\omega_1 + \omega_2 \neq 0$

$$F(x; \omega_1, L, T) = F(x; -\omega_1, L, T) = \overline{F(x; \omega_1, L, T)}$$

\Leftrightarrow The covariance \Leftrightarrow the power spectrum.

Bottom line: there's a natural description of the distribution of a time series, analogous to the variance of a scalar random variable

$$P_X(w) = \lim_{L \rightarrow \infty} \frac{1}{2} \left\langle \left| \int_0^L x(t) e^{-iwt} dt \right|^2 \right\rangle$$

- * Ests of $P_X(w)$ are (approximately) independent across w
[depends on estimation procedure]
- * $P_X(w)$ can be used to address questions on page ①.

$$* P_X(w) = \int_{-\infty}^{\infty} G_X(\tau) e^{-i\omega \tau} d\tau \quad \begin{bmatrix} \text{see} \\ \text{ps } ⑫ \text{ of} \\ \text{"Power Spectra"} \end{bmatrix}$$

- * $P_X(w)$ does not fully characterize X
(variable does not fully characterize a distribution)

(11)

Brief discussion for further extensions of the power spectrum to characterize noise:

A Consider $\langle F(x; \omega_1, L, T) F(x; \omega_2, L, T) F(x; \omega_3, L, T) \rangle$

analogous to \star in p. 9. Same argument shows that this = 0 unless

$$\omega_1 + \omega_2 + \omega_3 = 0.$$

So we can define the "h.spectrum"

$$B_X(\omega_1, \omega_2) = \lim_{L \rightarrow \infty} \frac{1}{L^{3/2}} \langle F(x; \omega_1, L, T) \cdot \\ F(x; \omega_2, L, T) \cdot \\ \overline{F(x; \omega_1 + \omega_2, L, T)} \rangle$$

Trispectrum, etc. can be similarly defined.

B Constructive proof that $P_X(\omega)$ does not characterize X fully:
(demo)

Given some $P_Y(\omega)$, construct a noise X for which

$$P_X(\omega) = P_Y(\omega).$$

Choose a large L . Randomly choose $\hat{x}(w)$ (indep. for each w)

with $\langle |\hat{x}(w)|^2 \rangle = L P_Y(w)$. Then create $x(t) = \int e^{iwt} \hat{x}(w) dw$

If we choose $\hat{x}(w)$ from a Gaussian distribution \mathcal{G} , then $x(t)$ will be Gaussian.

† need $\hat{x}(w) = \overline{\hat{x}(w)}$ to guarantee $x(t)$ real.

(12)

But we could have started with a noise Y which was not Gaussian, e.g.

$$y(t) = \begin{array}{c} \text{[square pulse]} \\ \dots \end{array}$$

What we do is to find a Gaussian noise X for which

$$P_X(\omega) = P_Y(\omega)$$

Y differs from X in higher moments.

When signals are combined/filtred, what happens to their power spectra?

Independent noises: power spectra add.

$$x(t) \xrightarrow{\quad} z(t) \qquad P_Z(\omega) = P_X(\omega) + P_Y(\omega)$$

because

$$F(z; \omega, L, T) = F(x; \omega, L, T) + F(y; \omega, L, T)$$

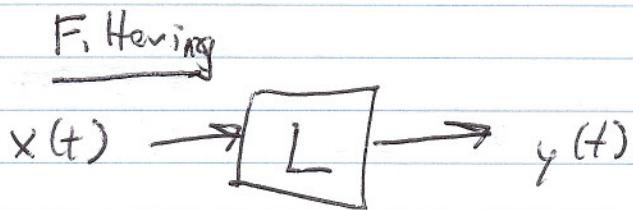
$$\begin{aligned} \text{so } \langle F(z) \bar{F(z)} \rangle &= \langle F(x) \bar{F(x)} \rangle + \langle F(x) \bar{F(y)} \rangle \\ &\quad + \langle F(y) \bar{F(x)} \rangle + \langle F(y) \bar{F(y)} \rangle \end{aligned}$$

Cross-terms = 0 because of independence.

NB: Say noises are not independent. Extreme case: $Y(t) = X(t)$
Then

$$P_Z(\omega) = 4P_X(\omega)$$

(13)



$$P_y(\omega) = |L(\omega)|^2 P_x(\omega).$$

This is because each Fourier estimate for Y ,

i.e., sonically of the form

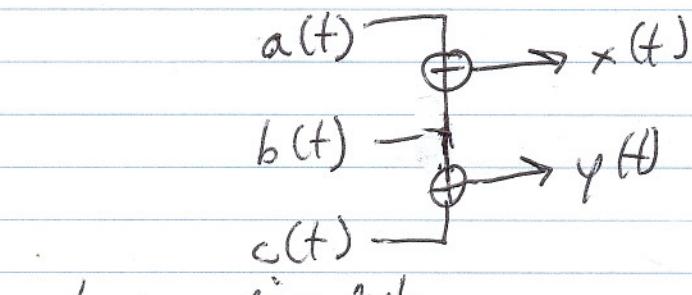
$$\int_{-T}^{L+T} y(t) e^{-i\omega t} dt,$$

simply writes y in the Fourier basis, and $\hat{y}_{\text{Four}} = \hat{L}(\omega) \hat{x}(\omega)$

i.e. the effect of L is to multiply an Fourier estimate of X

by $\hat{L}(\omega)$.

Common source



a, b, c independent.

How to analyze this? (Is it detect the common input?)

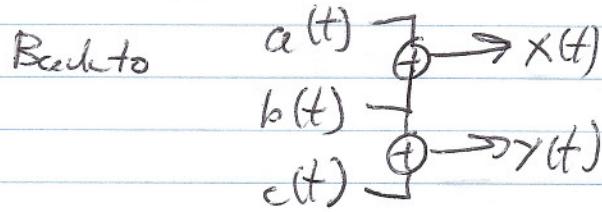
(1)

Time domain, - we'd look at $C_{XY}(\tau) = \langle X(t)Y(t+\tau) \rangle$.
 ("cross covariance")

By analogy to univariat case, we will instead look at

$$P_{XY}(w) = \lim_{L \rightarrow \infty} \frac{1}{L} \langle F(x; w, L, T) \overline{F(y; w, L, T)} \rangle.$$

This is the "cross-spectrum". Some comments apply that one needn't consider $\langle F(x; w) \overline{F(y; w)} \rangle$, and that estimates are independent (approximately) at different w .
 [But $P_{XY}(w)$ need not be real.]



$$P_X(w) = P_a(w) + P_b(w). \quad P_Y(w) = P_b(w) + P_c(w).$$

$$\begin{aligned} P_{XY}(w) &= \lim_{L \rightarrow \infty} \frac{1}{L} \langle (F(a; w, L, T) + F(b; w, L, T)) \\ &\quad (\overline{F(b; w, L, T) + F(c; w, L, T)}) \rangle \end{aligned}$$

$$= P_b(w), \text{ just what we want.}$$

Motivational def. of coherency = $\frac{P_{XY}(w)}{\sqrt{P_X(w) P_Y(w)}}$

\leftrightarrow coherence = | coherency |.

(15)

Cohärenz + cross spectra need not be real.

Say $y(t) = x(t - D)$, i.e., y is a delayed version of x .

$$\begin{aligned} \text{Then } F(y; \omega, \ell, T) &= \int_T^{T+L} y(t) e^{-j\omega t} dt \\ &= \int_T^{T+L} x(t-D) e^{-j\omega t} dt \\ &= \int_{T-D}^{T+2-D} x(\omega) e^{-j\omega(\omega+D)} d\omega \\ &\approx e^{-j\omega D} F(x; \omega, \ell, T-D) \end{aligned}$$

$$\begin{aligned} \text{so } \langle F(x; \omega, \ell, T) \overline{F(y; \omega, \ell, T)} \rangle \\ = e^{j\omega D} \langle F(x; \omega, \ell, T) \overline{F(x; \omega, \ell, T-D)} \rangle \end{aligned}$$

$$P_{XY}(\omega) = e^{j\omega D} P_X(\omega)$$

$$\text{Cohärenz} = e^{j\omega D}$$

$$\text{Cohärenz} = 1$$